

The above frequency-selective filtering is equivalent to the following operation in the time domain:

$$p(t) = \sum_{n=-\infty}^{\infty} p(n\Delta_t) \operatorname{sinc}\left[\frac{\pi(t - n\Delta_t)}{\Delta_t}\right].$$

This is called *reconstruction* or *interpolation* of the original signal from its delta-sampled signal.

The success of this operation hinges on the constraint

$$\omega_0 \leq \frac{2\pi}{2\Delta_t},$$

which is known as the *Nyquist* sampling criterion for *aliasing-free* or *error-free* reconstruction of  $p(t)$  from the sampled data  $p(n\Delta_t)$ 's. The linear harmonic components  $P(\omega - \frac{2\pi n}{\Delta_t})$ ,  $n = \pm 1, \pm 2, \dots$  (or  $n \neq 0$ ), result in aliasing if the Nyquist criterion is not satisfied.

The issue of sampling and its dual form, that is, sampling in the frequency domain, and aliasing-free processing and reconstruction from the sampled signal will be extensively encountered in array imaging systems.

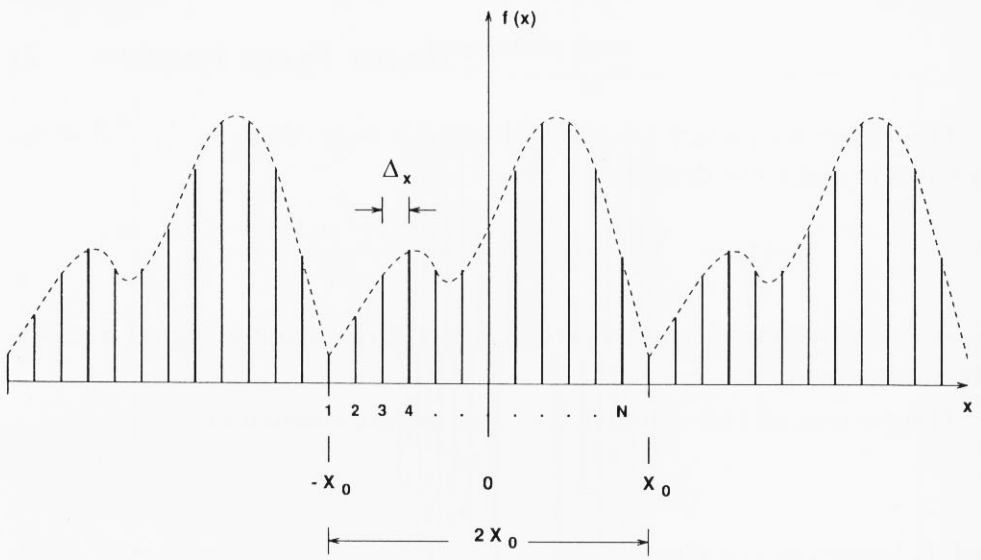
### 1.3 DISCRETE FOURIER TRANSFORM

Let  $f(x)$  be a periodic signal composed of evenly spaced delta functions, for example,

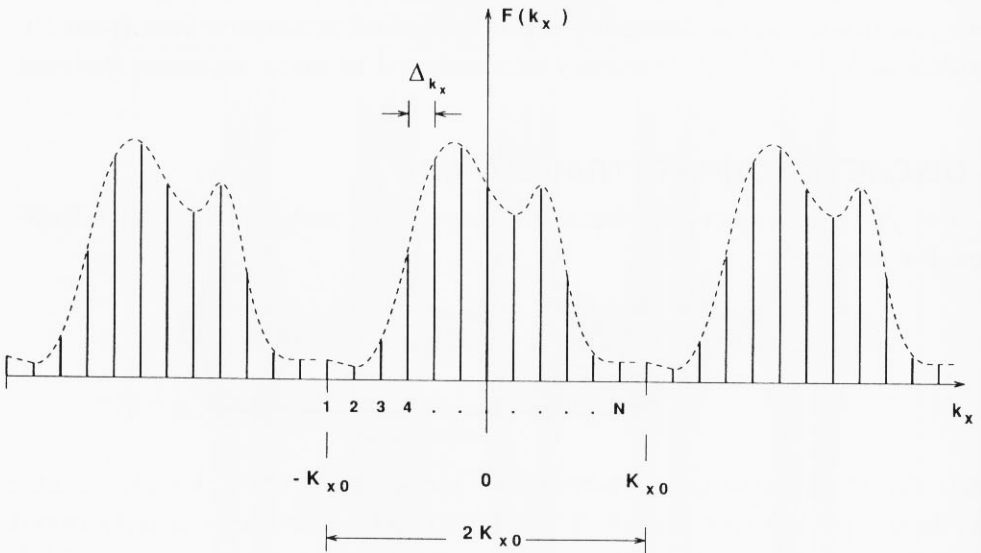
$$f(x) = \sum_{\ell=-\infty}^{\infty} \underbrace{\sum_{n=-N/2}^{N/2-1} f_n \delta[x - (n + N\ell)\Delta_x]}_{\text{One period}},$$

where  $\Delta_x$  is the spacing between two consecutive delta functions and  $N\Delta_x$  is the period (see Figure 1.7). This model provides a link between a discrete sequence and evenly spaced samples of a continuous signal [6]. Using the forward and inverse Fourier integrals, one can show that

$$F(k_x) = 2\pi \sum_{\ell=-\infty}^{\infty} \sum_{m=-N/2}^{N/2-1} F_m \delta[k_x - (m + N\ell)\Delta_{k_x}],$$



Depiction of a Discrete Fourier Transform Pair:  
The Spatial Domain Signal



Depiction of a Discrete Fourier Transform Pair:  
The Spatial Frequency Domain Signal

Figure 1.7 A discrete Fourier transform pair.

where

$$F_m = \sum_{n=-N/2}^{N/2-1} f_n \exp(-j \frac{2\pi}{N} mn) \quad (1.1)$$

$$f_n = \frac{1}{N} \sum_{m=-N/2}^{N/2-1} F_m \exp(j \frac{2\pi}{N} mn)$$

and

$$N \Delta_x \Delta_{k_x} = 2\pi. \quad (1.2)$$

Equations (1.1) and (1.2) are called the Discrete Fourier Transform (DFT) equations. It should be noted  $F(k_x)$  is also a periodic signal that is composed of evenly spaced delta functions (see Figure 1.7).

- *In practice, we deal with continuous functions that are not band-limited in space/time and frequency domains. For storage and processing purposes in a computer, we represent (approximate) them via periodic functions that are made up of a finite number of evenly spaced delta functions ( $N$  sampled data). The main period of  $f(x)$  [or  $F(k_x)$ ] is assumed to be the region in the  $x$  (or  $k_x$ ) domain, for example,  $x \in [-X_0, X_0)$  [or  $k_x \in [-K_{x0}, K_{x0})$ ], where most, for example, 95 percent, of its energy is concentrated ( $X_0$  and  $K_{x0}$  are known constants).  $[-X_0, X_0)$  and  $[-K_{x0}, K_{x0})$  are called the **effective support bandwidths** of  $f(x)$  and  $F(k_x)$ , respectively.*

Noting the fact that

$$2X_0 = N \Delta_x$$

$$2K_{x0} = N \Delta_{k_x},$$

we can write from (1.2) the following equations that are the Nyquist sampling rate (constraint) for representing/recovering a lowpass signal from its sampled data:

$$\Delta_x = \frac{\pi}{K_{x0}}$$

$$\Delta_{k_x} = \frac{\pi}{X_0}.$$