

EE 631: Estimation and Detection

Part 7

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Linear MMSE estimation

Consider an unknown message $m(t)$ which is estimated from the measured signal $r(t)$. Let the estimator be a linear system given by $h(t, u)$. The output of the estimator is given by:

$$\hat{m}(t) = \int_T h(t, u)r(u)du$$

Estimator error is given by:

$$e(t) = m(t) - \hat{m}(t)$$

Our aim is to design an optimum filter that minimizes

$$\xi_I = \int_T e^2(t)dt$$

and satisfies the orthogonality condition

$$E[e(t).r^*(u)] = 0 \quad \forall t, u \in [0, T]$$

Estimator equation

Substitute for the error in the orthogonality condition via:

$$\begin{aligned} e(t) &= m(t) - \hat{m}(t) \\ &= m(t) - \int h(t, \lambda)r(\lambda)d\lambda \end{aligned}$$

This yields:

$$\begin{aligned} E \left[[m(t) - \int_T h(t, \lambda)r(\lambda)d\lambda]r^*(u) \right] &= 0 \\ \Rightarrow E[m(t)r(u)] &= \int_T h(t, \lambda)E[r(\lambda)r^*(u)]d\lambda \\ \Rightarrow R_{mr}(t, u) &= \int_T h(t, \lambda)R_r(\lambda, u)d\lambda \end{aligned}$$

The construction of the filter is based on solving for $h(t, u)$ from the above using the knowledge of $R_{mr}(\cdot, \cdot)$ and $R_r(\cdot, \cdot)$.

Special case:

$$r(t) = m(t) + n(t)$$

where $m(t)$ and $n(t)$ are orthogonal to each other. In this case:

$$\begin{aligned} R_{mr}(t, u) &= E \left[m(t)r^*(u) \right] \\ &= E \left[m(t)[m^*(u) + n^*(u)] \right] \\ &= R_m(t, u) \end{aligned}$$

Also

$$R_r(t, u) = R_m(t, u) + R_n(t, u)$$

Therefore, the estimator equation becomes:

$$R_m(t, u) = \int_T h(t, \lambda)[R_m(\lambda, u) + R_n(\lambda, u)]d\lambda$$

Spectral solution via KL representation

Model:

$$r(t) = m(t) + n(t)$$

where $m(t)$ and $n(t)$ are general non-stationary signals that are orthogonal and uncorrelated to each other. Suppose the autocorrelation of both signals possess a common set of eigenfunctions that are denoted by:

$$\Phi = \{\phi_i(t); \quad ; i = 1, 2, \dots\}$$

Thus,

$$R_m(t, u) = \sum_i \lambda_{mi} \phi_i(t) \phi_i^*(u)$$

$$R_n(t, u) = \sum_i \lambda_{ni} \phi_i(t) \phi_i^*(u)$$

We consider the following decomposition for the optimal filter:

$$h(t, u) = \sum_i h_i \phi_i(t) \phi_i^*(u)$$

Also, we have:

$$\begin{aligned} R_r(t, u) &= \sum_i \lambda_{ri} \phi_i(t) \phi_i^*(u) \\ &= \sum_i (\lambda_{mi} + \lambda_{ni}) \phi_i(t) \phi_i^*(u) \end{aligned}$$

We now use the spectral representations in the estimator equation:

$$\begin{aligned} R_m(t, u) &= \int_T h(t, \lambda) R_r(\lambda, u) d\lambda \\ \sum \lambda_{mi} \phi_i(t) \phi_i^*(u) &= \int_T \sum_i h_i \phi_i(t) \phi_i^*(\lambda) \sum_j (\lambda_{mj} + \lambda_{nj}) \phi_j(\lambda) \phi_j^*(u) d\lambda \\ &= \sum_i \sum_j h_i (\lambda_{mj} + \lambda_{nj}) \phi_i(t) \phi_j^*(u) \underbrace{\left[\int_T \phi_i^*(\lambda) \phi_j(\lambda) d\lambda \right]}_{\delta_{ij}} \\ \therefore \sum_i \lambda_{mi} \phi_i(t) \phi_i^*(u) &= \sum_i h_i (\lambda_{mi} + \lambda_{ni}) \phi_i(t) \phi_i^*(u) \end{aligned}$$

Due to the uniqueness of the spectral coefficients,

$$\begin{aligned} \lambda_{mi} &= h_i (\lambda_{mi} + \lambda_{ni}) \quad \forall \quad i = 1, 2, \dots \\ \Rightarrow h_i &= \frac{\lambda_{mi}}{\lambda_{mi} + \lambda_{ni}} \end{aligned}$$

which is known as the general Weiner filter. The stationary equivalent of the above is given by:

$$H(\omega) = \frac{S_m(\omega)}{S_m(\omega) + S_n(\omega)}$$

Note that:

a) $\lambda_{mi} \gg \lambda_{ni} \Rightarrow h_i \approx 1$

b) $\lambda_{mi} \ll \lambda_{ni} \Rightarrow h_i \approx 0$

The estimate:

The estimate is constructed via the following:

$$\hat{m}(t) = \int_T h(t, u) r(u) du$$

We consider the spectral decomposition for the measurement:

$$\begin{aligned} m(t) &= \sum_i m_i \phi_i(t) \\ n(t) &= \sum_i n_i \phi_i(t) \\ \Rightarrow r(t) &= \sum_i r_i \phi_i(t) \quad ; r_i = m_i + n_i \end{aligned}$$

Using this and the spectral representation of for $h(t, u)$ in $\hat{m}(t)$, we get:

$$\begin{aligned} \hat{m}(t) &= \int_T \sum_i h_i \phi_i(t) \phi_i^*(u) \sum_j r_j \phi_j(u) du \\ &= \sum_i \sum_j h_i r_j \phi_i(t) \underbrace{\left[\int_T \phi_i^*(u) \phi_j(u) du \right]}_{\delta_{ij}} \\ &= \sum_i h_i r_i \phi_i(t) \\ \rightarrow \hat{m}(t) &= \sum_i \hat{m}_i \phi_i(t) \end{aligned}$$

where $\hat{m}_i = h_i r_i = \frac{\lambda_{m_i} r_i}{\lambda_{m_i} + \lambda_{n_i}}$

Estimator error energy:

Point error:

$$\begin{aligned} \xi_p(t) &= E \left[[m(t) - \hat{m}(t)][m^*(t) - \hat{m}^*(t)] \right] \\ &= E \left[e(t)[m^*(t) - \int_T h(t, u) r^*(u) du] \right] \\ &= E[e(t)m^*(t)] - \int_T h(t, u) \underbrace{E[e(t)r^*(u)]}_{=0 \text{ due to orthogonality}} du \\ \Rightarrow \xi_p(t) &= E[e(t)m^*(t)] \\ &= E \left[[m(t) - \hat{m}(t)]m^*(t) \right] \\ &= R_m(t, t) - E \left[\int_T h(t, u) r(u) du m^*(t) \right] \\ &= R_m(t, t) - \int_T h(t, u) \underbrace{E[r(u)m^*(t)]}_{=R_m(u, t)} du \\ &= R_m(t, t) - \int_T h(t, u) R_m(u, t) du \\ &= R_m(t, t) - \int_T \sum_i h_i \phi_i(t) \phi_i^*(u) \sum_j \lambda_{m_j} \phi_j(u) \phi_j^*(t) du \\ &= R_m(t, t) - \sum_i \sum_j h_i \lambda_{m_j} \phi_i(t) \phi_j^*(t) \underbrace{\left[\int_T \phi_i^*(u) \phi_j(u) du \right]}_{\delta_{ij}} \end{aligned}$$

$$\begin{aligned}
\Rightarrow \xi_p(t) &= \sum_i \lambda_{mi} \phi_i(t) \phi_i^*(t) - \sum_i h_i \lambda_{mi} \phi_i(t) \phi_i^*(t) \\
&= \sum_i \left(\lambda_{mi} - \frac{\lambda_{mi}}{\lambda_{mi} + \lambda_{ni}} \lambda_{mi} \right) \phi_i(t) \phi_i^*(t) \\
&= \sum_i \left(\frac{\lambda_{mi} \lambda_{ni}}{\lambda_{mi} + \lambda_{ni}} \right) |\phi_i(t)|^2
\end{aligned}$$

Interval error:

$$\begin{aligned}
\xi_I &= \int_T \xi_p(t) dt \\
&= \int_T \sum_i \left(\frac{\lambda_{mi} \lambda_{ni}}{\lambda_{mi} + \lambda_{ni}} \right) |\phi_i(t)|^2 dt \\
&= \sum_i \left(\frac{\lambda_{mi} \lambda_{ni}}{\lambda_{mi} + \lambda_{ni}} \right) \underbrace{\int_T |\phi_i(t)|^2 dt}_{=1} \\
\Rightarrow \xi_I &= \sum_i \left(\frac{\lambda_{mi} \lambda_{ni}}{\lambda_{mi} + \lambda_{ni}} \right)
\end{aligned}$$

Detection in additive white Gaussian noise

Under the m th hypothesis, the received signal is :

$$H_m : \quad r(t) = s_m(t) + n(t)$$

where $n(t)$ is the white Gaussian noise and the transmitted signals are:

$$\vec{S} \triangleq \{s_m(t); \quad m = 0, 1, \dots, M-1\}$$

The linear signal subspace can be represented by (e.g using the Gram Schmidt procedure) by $N \leq M$ orthonormal basis functions given by:

$$\vec{S} = \vec{\Phi} = \{\phi_i(t); \quad i = 1, \dots, N \leq M\}$$

Irrelevant data:

The finite signal subspace $\vec{S} = \vec{\Phi}$ cannot be a CON set. Therefore, we identify the set:

$$\vec{S}^c \triangleq \{\phi_i(t); \quad i = N+1, \dots, \infty\}$$

such that $[\vec{S}, \vec{S}^c]$ forms a CON set.

Once this signal set is identified, we can construct:

1. Projection of $S_m(t)$ into \vec{S} :

$$\vec{S}_m = \begin{bmatrix} S_{m1} \\ S_{m2} \\ \vdots \\ S_{mN} \end{bmatrix}$$

where $S_{mi} = \langle S_m, \phi_i \rangle \quad \forall \quad i = 1, 2, \dots, N.$

Clearly, $\langle S_m, \phi_i \rangle = 0 \quad \forall \quad i = N+1, \dots, \infty$

2. Projection of $r(t)$ into \vec{S} :

$$\vec{R}_1 = \begin{bmatrix} r_1 \\ r_2 \\ \vdots \\ r_N \end{bmatrix}$$

where $r_i = \langle R, \phi_i \rangle \quad \forall i = 1, 2, \dots, N$.

Under H_m : $r_i = S_{mi} + n_i \quad ; i = 1, 2, \dots, N$,

where $n_i = \langle n, \phi_i \rangle$.

Furthermore, $\forall i = N + 1, \dots, \infty$, we also have $r_i = n_i$, i.e. no signal component.

$$\vec{R}_2 = \begin{bmatrix} r_{N+1} \\ r_{N+2} \\ \vdots \\ r_\infty \end{bmatrix}$$

Therefore, the total measurement vector is given by:

$$\vec{R} = \begin{bmatrix} r_1 \\ r_2 \\ \vdots \\ r_N \\ r_{N+1} \\ r_{N+2} \\ \vdots \\ r_\infty \end{bmatrix} = \begin{bmatrix} R_1 \\ R_2 \end{bmatrix}$$

Under H_m :

$$\vec{R} = \begin{bmatrix} s_{m1} + n_1 \\ s_{m2} + n_2 \\ \vdots \\ s_{mN} + n_N \\ n_{N+1} \\ n_{N+2} \\ \vdots \\ n_\infty \end{bmatrix}$$

Since R_2 has only noise components, it is called *redundant data*.

Noise variance:

$$E[|n_i|^2] = \frac{N_0}{2}$$

$$E[n_i n_j^*] = 0 \quad ; i \neq j$$

$\Rightarrow n_i$'s are i.i.d. $\sim N(0, \frac{N_0}{2})$. R_1 and R_2 are independent of each other. i.e.

$$p(\vec{R}|H_m) = p(\vec{R}_1|H_m)p(\vec{R}_2|H_m)$$

Since, R_2 does not depend on H_m ,

$$p(\vec{R}_2|H_m) = p(R_2)$$

The likelihood ratio test:

$$\Lambda_m(\vec{R}) = \frac{p(\vec{R}|H_m)}{p(\vec{R}|H_0)} = \frac{p(\vec{R}_1|H_m)}{p(\vec{R}_1|H_0)}$$

i.e. \vec{R}_1 is sufficient statistic. R_2 on the other hand is irrelevant data and does not influence decision.

Statistic on R_1 :

R_1 is multivariate normal. Under H_m , its mean is S_m and its covariance matrix is $\frac{N_0}{2} \times \vec{I}_{N \times N}$.